

MASTER OSCILLATOR

The frequency of the master oscillator is determined by the contents of several output ports. Port 10H sets the master frequency. It is given by the following formula:

$$F_m = \frac{1789}{\text{PORT } 10H + 1} \text{ KHz}$$

If bit 4 of output port 15H is set to 1, the master oscillator frequency will be modulated by noise. The amount of modulation will be set by the 8-bit noise volume register, output port 17H.

If bit 4 of output port 15H is set to 0, the frequency of the master oscillator will be modulated by a constant value to give a vibrato effect. The amount of modulation will be set by the vibrato depth register (the first 6 bits of output port 14H). The speed of modulation is set by the vibrato speed register (upper 2 bits of output port 14H); 00 for fastest and 11 for slowest.

Frequency modulation is accomplished by adding a modulation value to the contents of port 10H and sending the result to the master oscillator frequency generator. In noise modulation, the modulation value is an 8-bit word from the noise generator. If a bit in the noise volume register is set to 1, the corresponding bit in the modulation value word will be set to 1. In vibrato modulation, the modulation value alternates between 0 and the contents of the vibrato volume register.

Modulation can be completely disabled by setting the master volume to 0 if noise modulation is being used, or by setting the vibrato depth to 0 when vibrato is used.

TONES

The system contains three tone generators each clocked by the same master oscillator. The frequency of Tone A is set by output port 11H, Tone B by output port 12H, and Tone C by output port 13H. The frequency is given by the following formula:

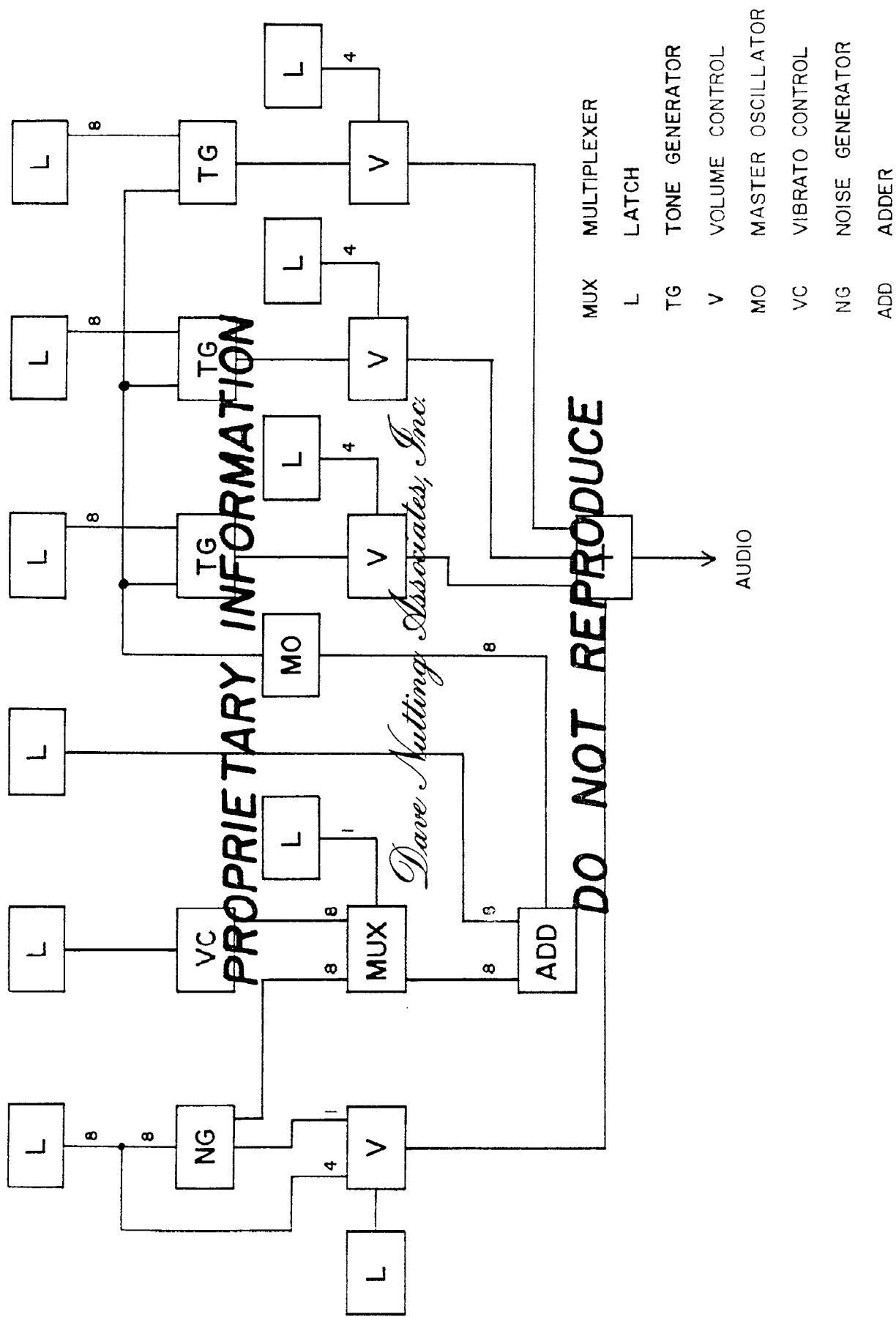
$$F_f = \frac{F_m}{2(\text{contents of TONE PORT} + 1)} = \frac{894}{(\text{PORT } 10H + 1)(\text{contents of TONE PORT} + 1)} \text{ KHz}$$

The tone volumes are controlled by output ports 15H and 16H. The lower 4 bits of port 16H set Tone A Volume, the upper 4 bits sets Tone B Volume. The lower 4 bits of port 15H sets Tone C Volume. Noise can be mixed with the tones by setting bit 5 of port 15H to 1. In this case the noise volume is given by the upper 4 bits of port 17H. In all cases a volume of 0 is silence and a volume of all 1's is loudest.

SOUND BLOCK TRANSFER

All 8 bytes of sound control can be sent to the audio circuit with one OTIR instruction. Register C should be sent to 18H, register B to 8H and HL pointing to the 8 bytes of data. The data pointed to by HL goes to port 17H and the next 7 bytes of data goes to ports 16H through 10H.

HL →	Memory Location	X	Data-to-port	17H
		X+1	Data-to-port	16H
		X+2	Data-to-port	15H
		X+3	Data-to-port	14H
		X+4	Data-to-port	13H
		X+5	Data-to-port	12H
		X+6	Data-to-port	11H
		X+7	Data-to-port	10H



AUDIO GENERATOR BLOCK DIAGRAM

OUTPUT PORTS

<u>PORT NUMBER</u>	<u>FUNCTION</u>
0H	Color Register 0
1H	Color Register 1
2H	Color Register 2
3H	Color Register 3
4H	Color Register 4
5H	Color Register 5
6H	Color Register 6
7H	Color Register 7
8H	Low/High Resolution
9H	Horizontal Color Boundary, Background Color
AH	Vertical Blank Register
BH	Color Block Transfer
CH	Magic Register
DH	Interrupt Feedback Register
EH	Interrupt Enable and Mode
FH	Interrupt Line
10H	Master Oscillator
11H	Tone A Frequency
12H	Tone B Frequency
13H	Tone C Frequency
14H	Vibrato Register
15H	Tone C Volume, Noise Modulation Control
16H	Tone A Volume, Tone B Volume
17H	Noise Volume Register
18H	Sound Block Transfer
19H	Expand Register

PROPRIETARY INFORMATION**DO NOT REPRODUCE**

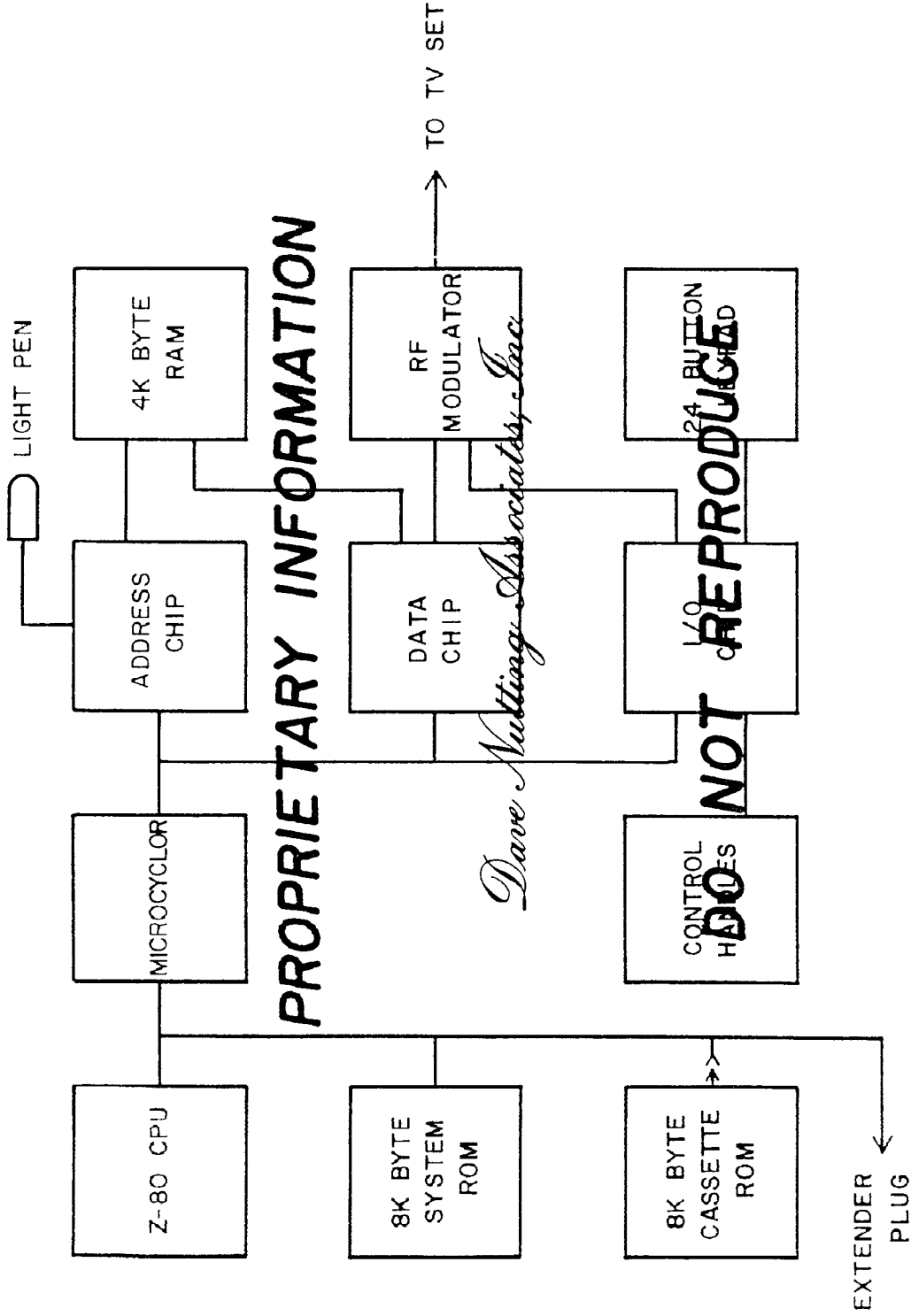
INPUT PORTS

<u>PORT NUMBER</u>	<u>FUNCTION</u>
8H	Intercept Feedback
EH	Vertical Line Feedback
FH	Horizontal Address Feedback
10H	Player 1 Handle
11H	Player 2 Handle
12H	Player 3 Handle
13H	Player 4 Handle
14H	Keypad Column 0 (right)
15H	Keypad Column 1
16H	Keypad Column 2
17H	Keypad Column 3 (left)

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SYSTEM BLOCK DIAGRAM

MICROCYCLER

The purpose of the microcycler is to combine the 16-bit Address Bus and the 8-bit Data Bus from the Z-80 into one 8-bit Microcycle Data Bus to the Data Chip, Address Chip, and I/O Chip. This was done to reduce the pin count on the custom chips.

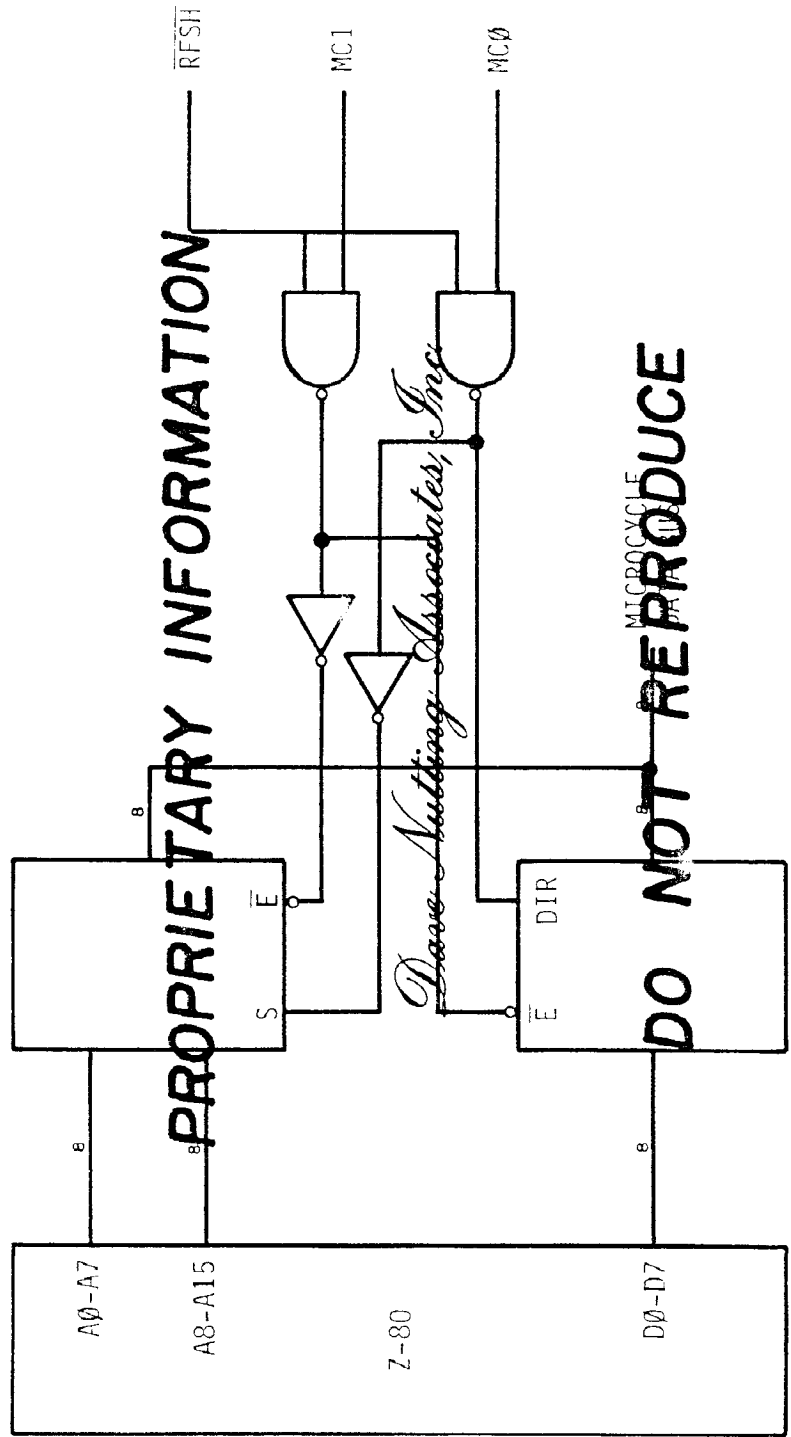
The Microcycle Data Bus can be in any of four modes. Its mode is controlled by MC0 and MC1 coming from the Data Chip and RFSH from the Z-80. The modes are shown below.

<u>RFSH</u>	<u>MC0</u>	<u>MC1</u>	<u>Microcycle Data Bus Contents</u>
0	0	0	A0 - A7 from Z-80
0	0	1	A0 - A7 from Z-80
0	1	0	A0 - A7 from Z-80
0	1	1	A0 - A7 from Z-80
1	0	0	A0 - A7 from Z-80
1	0	1	A8 - A15 from Z-80
1	1	0	D0 - D7 from Z-80
1	1	1	D0 - D7 to Z-80

MC0 and MC1 change 140 nsec after the rising edge of ϕ . Their changes are shown in the timing diagrams of various instruction cycles.

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MICROCYCLER BLOCK DIAGRAM

ADDRESS CHIP DESCRIPTION

The Microcycle Decoder generates twelve bits of Z-80 address from the 8-bit Microcycle Data Bus. This address is then fed through MUX I and MUX II to MA0-5 which go to the RAM. The Scan Address Generator generates a 12-bit address which is used to read video data from the RAM. This address goes from 0 to FFFH once every frame (1/60 sec.).

MUX I sends either the Scan Address or Z-80 Address to its 12 outputs. An output of the Scan Address Generator controls MUX I. If the Scan Address Generator and the Z-80 request a memory cycle at the same time, the Scan Address Generator will have higher priority and the Z-80 will be required to wait (by the $\overline{\text{WAIT}}$ output). The Scan Address Generator never requires the memory for more than one consecutive memory cycle, so the Z-80 is never required to wait for the memory for more than one cycle. HORIZ DR and VERT DR synchronize the Scan Address Generator with the Data Chip and the TV Scan.

The purpose of MUX II is to multiplex its 12 inputs to the six address bits in the two time slots required for 4K x 16 pin RAMS.

The Memory Cycle Generator controls memory cycles generated by either the Z-80 or Scan Address Generator. $\overline{\text{MREQ}}$, $\overline{\text{RD}}$, $\overline{\text{MI}}$, $\overline{\text{RFSH}}$, and A12-A15 are from the Z-80. A12-A15 are fed directly from the Z-80 because if they were brought out of the microcycle decoder, they would arrive too late in the memory cycle. The RAS0 - RAS3 outputs are used to activate memory cycles. In the consumer game, only RAS0 is used to one bank of RAM (4K x 8). In the commercial game, all four RAS's are used to control four banks of RAM (16K x 8). WRCTL and LTCHDO are control signals to the Data Chip. WRCTL tells the Data Chip when to place data to be written to memory on the memory data bus. LTCHDO tells the Data Chip when valid data from RAM is present on the memory data bus.

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As mentioned earlier, $\overline{\text{WAIT}}$ is generated when the Z-80 and Scan Address Generator both request memory at the same time. $\overline{\text{WAIT}}$ is also generated for one cycle every time the Z-80 requests a memory access, even if there is no conflict with the Scan Address. This is because the microcycle slows down Z-80 memory accesses. The Z-80 address bus and data bus must time share the microcycle bus so the Z-80 data reaches the microcycle bus very late in the memory cycle.

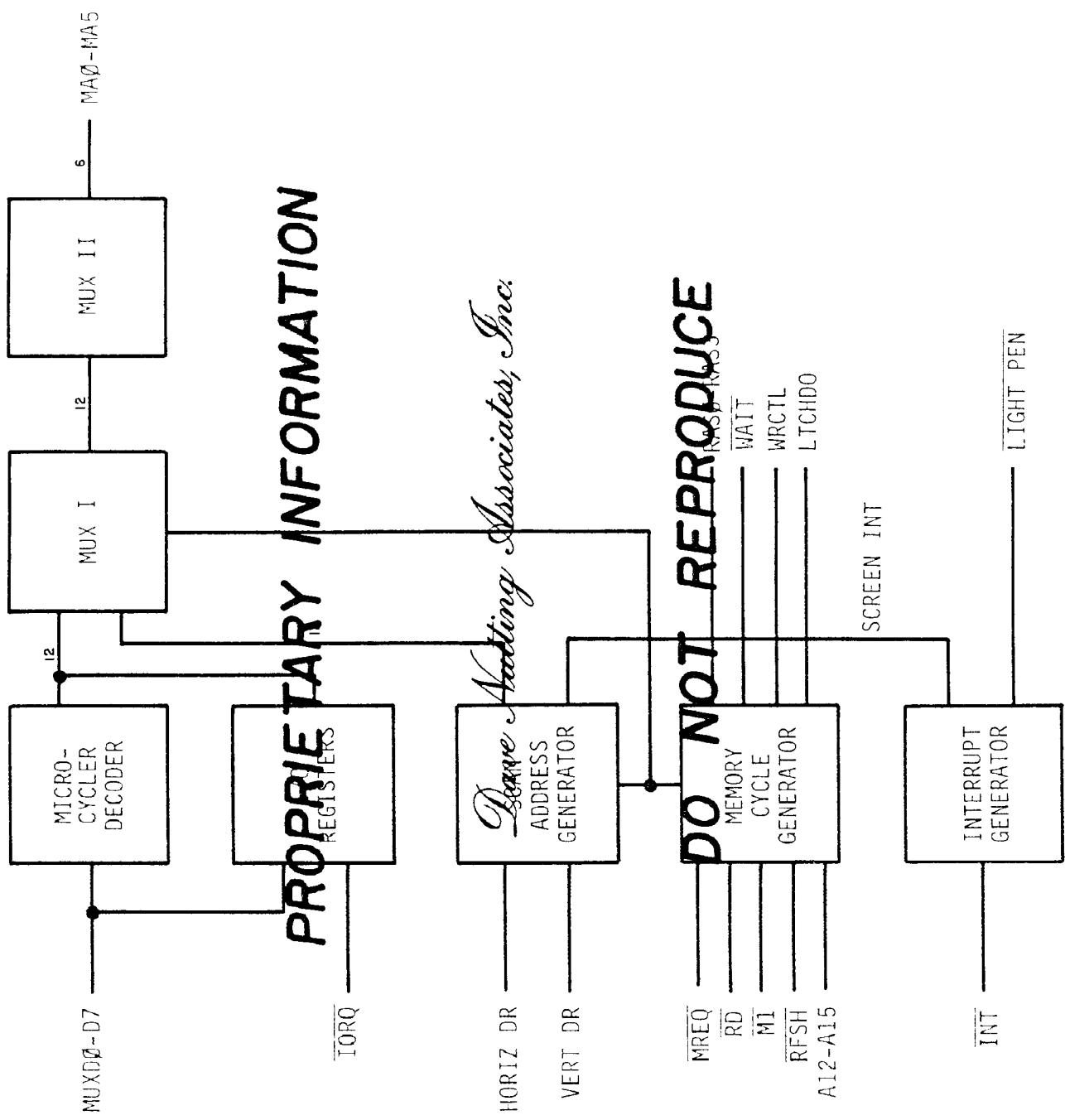
The INT Generator generates two types of interrupts for the Z-80; Light Pen and Screen interrupts. A screen interrupt is generated when screen interrupts are enabled and the TV scan completes a certain line on the screen (from 0 to 255). The line at which the interrupt will occur is determined by the Z-80. This interrupt can be used for timing since the TV rescans every line once every 1/60 s. A light pen interrupt occurs when the light pen interrupt is enabled and $\overline{\text{LIGHT PEN}}$ goes low. The current scan address is saved in latches in the Scan Address Generator. The Z-80 can read the contents of these latches to determine the scan address at the time $\overline{\text{LIGHT PEN}}$ was activated and thus the position of the light pen on the screen.

The I/O Decode circuit is used during Z-80 input and output instructions. Z-80 input instructions are used to read the scan address after light pen interrupts. Output instructions are used to enable the two interrupts and set the line number for screen interrupts.

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ADDRESS CHIP BLOCK DIAGRAM

DATA CHIP DESCRIPTION

The TV Sync Generator uses $7M$ and $\overline{7M}$ (7.159090 Mhz square waves) to generate NTSC standard sync and blank to be sent to the Video Generator. It also generates HORIZ DR and VERT DR for synchronization with the Address Chip. HORIZ DR occurs once every horizontal line (63.5 usec), and VERT DR occurs once every frame (16.6 msec).

The Shift Register loads parallel data from the memory data bus (MD \emptyset - MD7) and shifts it out of it two serial outputs. The TV Sync Generator controls when data is loaded or shifted. In a consumer game, the two outputs of the shift register are sent through MUX I to MUX II. In a commercial game, SERIAL \emptyset and SERIAL 1 are sent through the MUX I to MUX II. The two bits from MUX I select 8 bits to be sent through MUX II to the Video Generator. These 8 bits then determine the analog values of VIDEO, R-Y, and B-Y. 2.5V is a 2.5V D C reference level.

The Clock Generator generates $\emptyset G$ and PX from $7M$. These are the clocks for the rest of the system. The frequency of \overline{PX} is half that of $7M$ and the frequency of $\emptyset G$ is half that of \overline{PX} .

The Microcycle Generator generates the microcycle control bits, MC \emptyset and MC1, from \overline{IORQ} , \overline{MREQ} , P_0 , and $\overline{M1}$, all from the Z-8 \emptyset .

In memory write cycles WRCTL is activated and the Memory Control circuit generates \overline{DATEN} . The Magic Function Generator takes the data from the Z-8 \emptyset on MUXD \emptyset - D7 and transfers it to MD \emptyset - MD7. If a Magic write is being done, the Magic Function Generator will modify the data as required before it places it on the memory data bus.

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A Magic write is a memory write cycle in which data is written to a location, (X) from 0 to 16K. All memory from 0 to 16K is ROM and cannot be modified. The data is modified by the Magic Function Generator and is written to location X + 16K. The way in which the data is modified is determined by the 7 bits coming from the I/O registers.

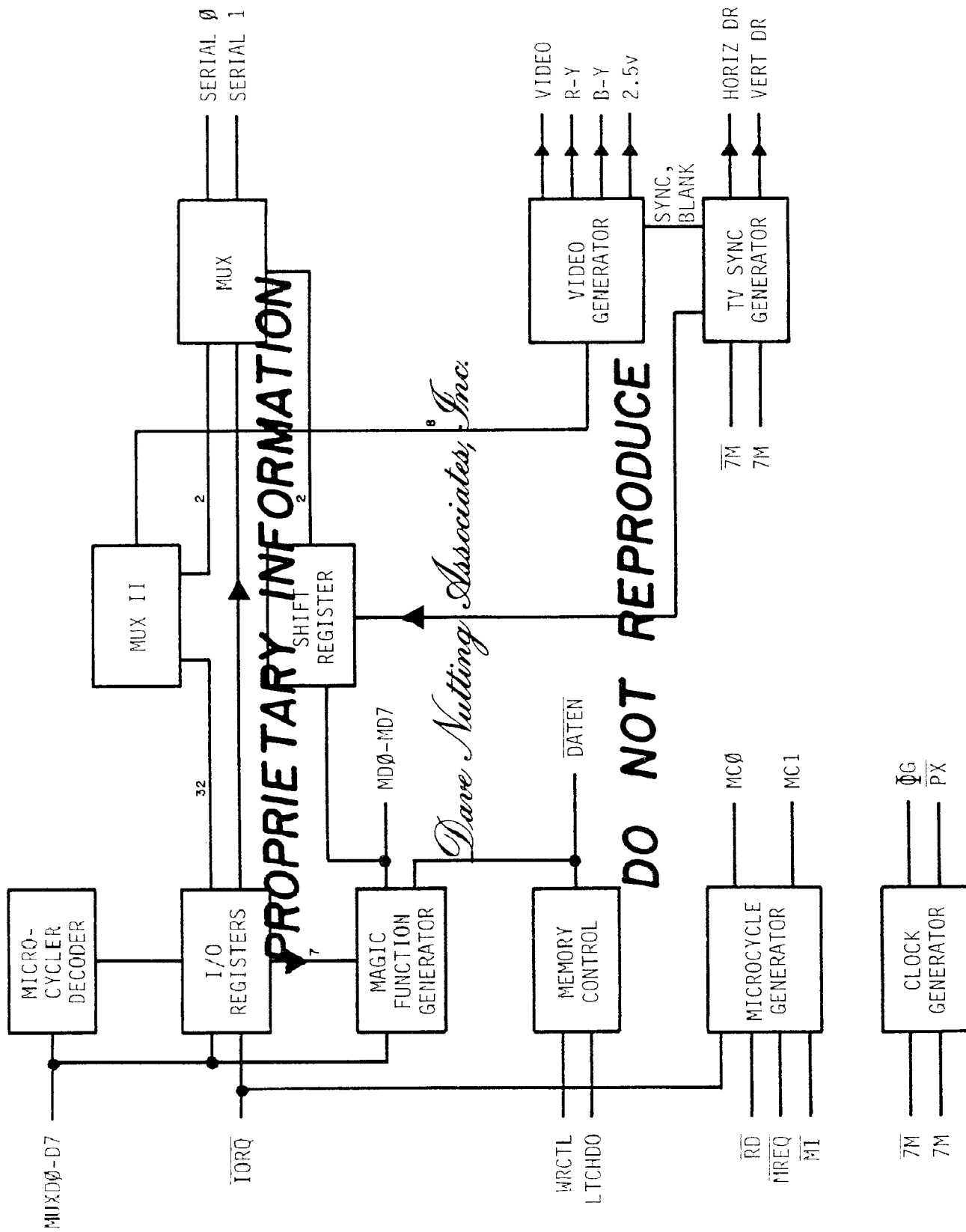
In memory reads, data is transferred from MD0 - MD7 to MUXD0 - MUXD7. Also, LTCHDO is activated which causes the data from RAM to be latched up in a register in the Magic Function Generator. This latched data is used in some magic functions.

The I/O registers are loaded by output instructions from the Z-80 just as in the Address Chip.

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DATA CHIP BLOCK DIAGRAM

I/O CHIP DESCRIPTION

The Z-80 communicates with the I/O Chip through input and output instructions. The state of an 8 x 8 switch matrix can be read through the Switch Scan circuit. When an input instruction is executed, one of the S00-S07 lines will be activated. When a line is activated, the switch matrix will feed back eight bits of data on SI0-SI7. This data is in turn fed to the Z-80 through MUXD0 - MUXD7.

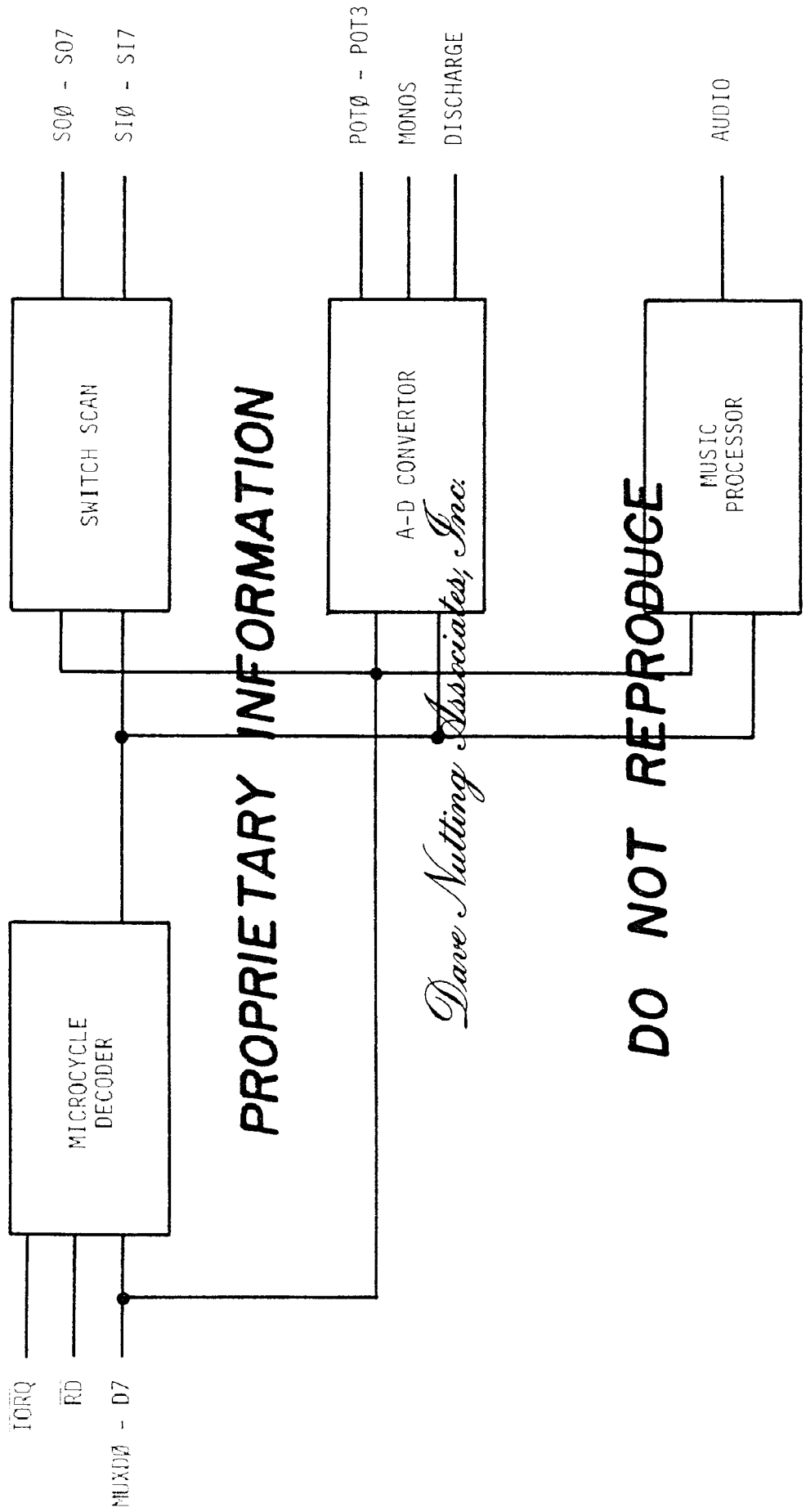
The Z-80 can read the position of four potentiometers (pots) through the A-D Converter circuit. The pots are continuously scanned by the A-D Converter and the results of the conversions are stored in a RAM in the A-D Converter circuit. The Z-80 simply reads this RAM with input instructions.

The Z-80 loads data into the Music Processor with output instructions. This data determines the characteristics of the audio that is generated. The Music Processor is described in detail below.

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I/O CHIP BLOCK DIAGRAM

MUSIC PROCESSOR

The music processor can be divided into two sections. The first section generates the Master Oscillator Frequency and the second section uses the Master Oscillator Frequency to generate tone frequencies and the analog audio output. The contents of all registers in the Music Processor are set by output instructions from the Z-80.

Master Oscillator Frequency is a square wave whose frequency is determined by the 8 binary inputs to the Master Oscillator. The 8-bit word is the sum of the contents of the Master Oscillator Register and the output of the MUX. The MUX is controlled by MUX REG.

If MUX REG contains 0, then data from the Vibrato System will be fed through the MUX. The two bits from the Vibrato Frequency Register determine the frequency of the square wave output of the Low Frequency Oscillator. The 6-bit word at the output of the AND gates oscillates between 0 and the contents of the Vibrato Register. The frequency of oscillation is determined by the contents of the Vibrato Frequency Register. The 6-bit word, along with two ground bits are fed through the MUX to the Adder. This causes the Master Oscillator Frequency to be modulated between two values thus giving a vibrato effect.

If MUX REG contains 1, then data from the Noise System will be fed through the MUX. The 8-bit word from the Noise Volume Register determines which bits from the Noise Generator will be present at the output of the AND gates.

If a bit in the Noise Volume Register is 0, then the corresponding bit at the output of the AND gates will be 0. If a bit in the Noise Volume Register is 1, then the corresponding bit at the output of the AND gates will be noise from the Noise Generator. This 8-bit word is sent through the MUX to the Adder. The Master Oscillator Frequency is modulated by noise.

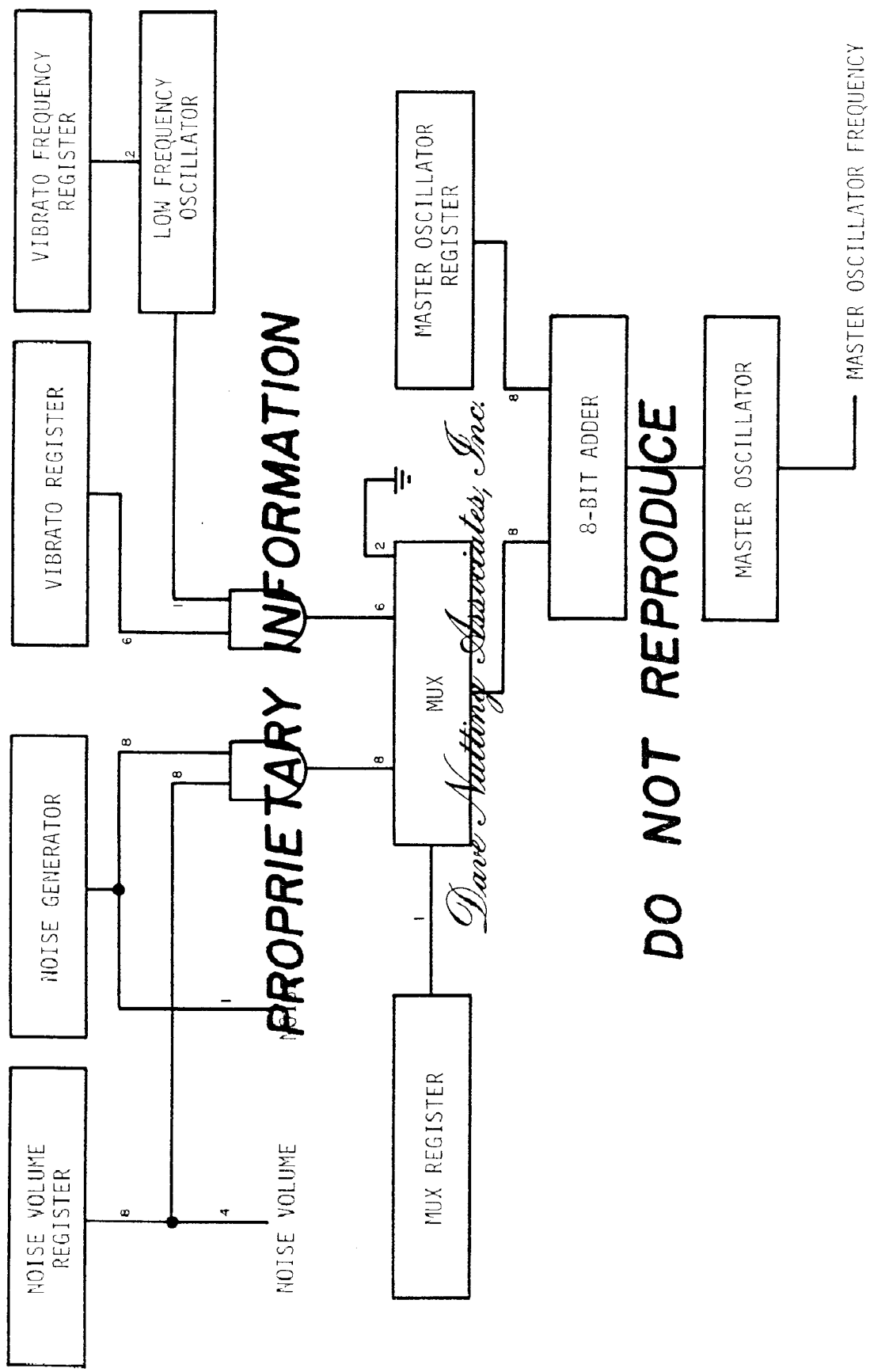
In the second part of the Music Processor, the square wave from the Master Oscillator is fed to three Tone Generator circuits which produce square waves at their outputs. The frequency of their outputs is determined by the contents of their Tone Generator Register and Master Oscillator Frequency. The 4-bit words at the output of the AND gates oscillate between 0 and the contents of the Tone Volume Register. These 4-bit words are sent to D-A Converters whose outputs oscillate between GND and a positive analog voltage determined by the contents of the Tone Volume Register.

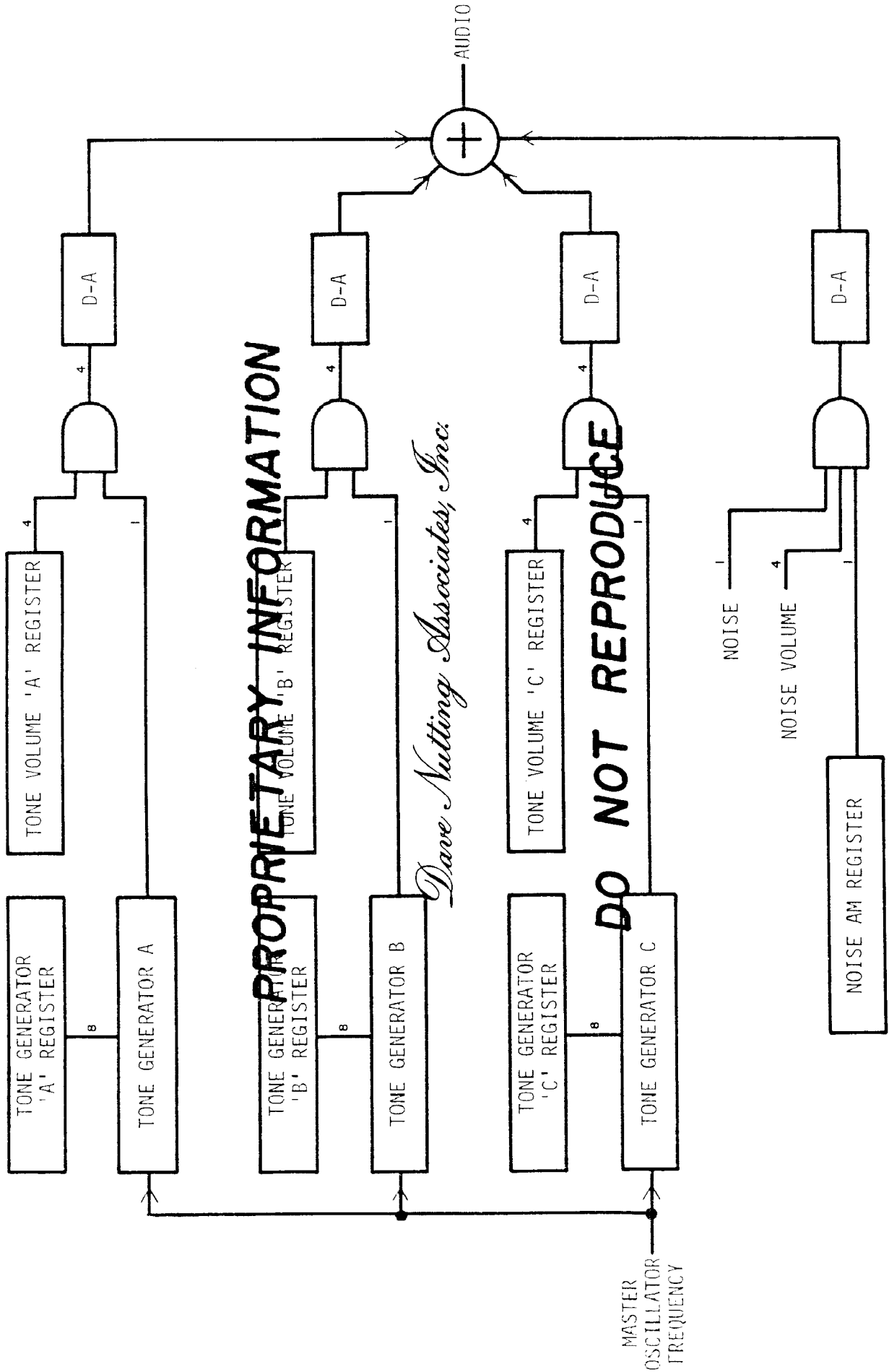
One Noise bit and four Noise Volume bits from the first section of the Music Processor are fed to a set of AND gates. This set of AND gates operates the same way as the AND gates for the tones except that the Noise AM Register must contain a 1 for the outputs of the AND gates to oscillate. The analog outputs of the four D-A Converters are summed to produce the single audio output.

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CUSTOM CHIP TIMING

The following diagrams show the relationship of various signals in the system during different types of operations. Delays are shown to be zero nsec from the clock edge which causes the transition. The actual delay is given in "Electrical Specification for Midway Custom Circuits".

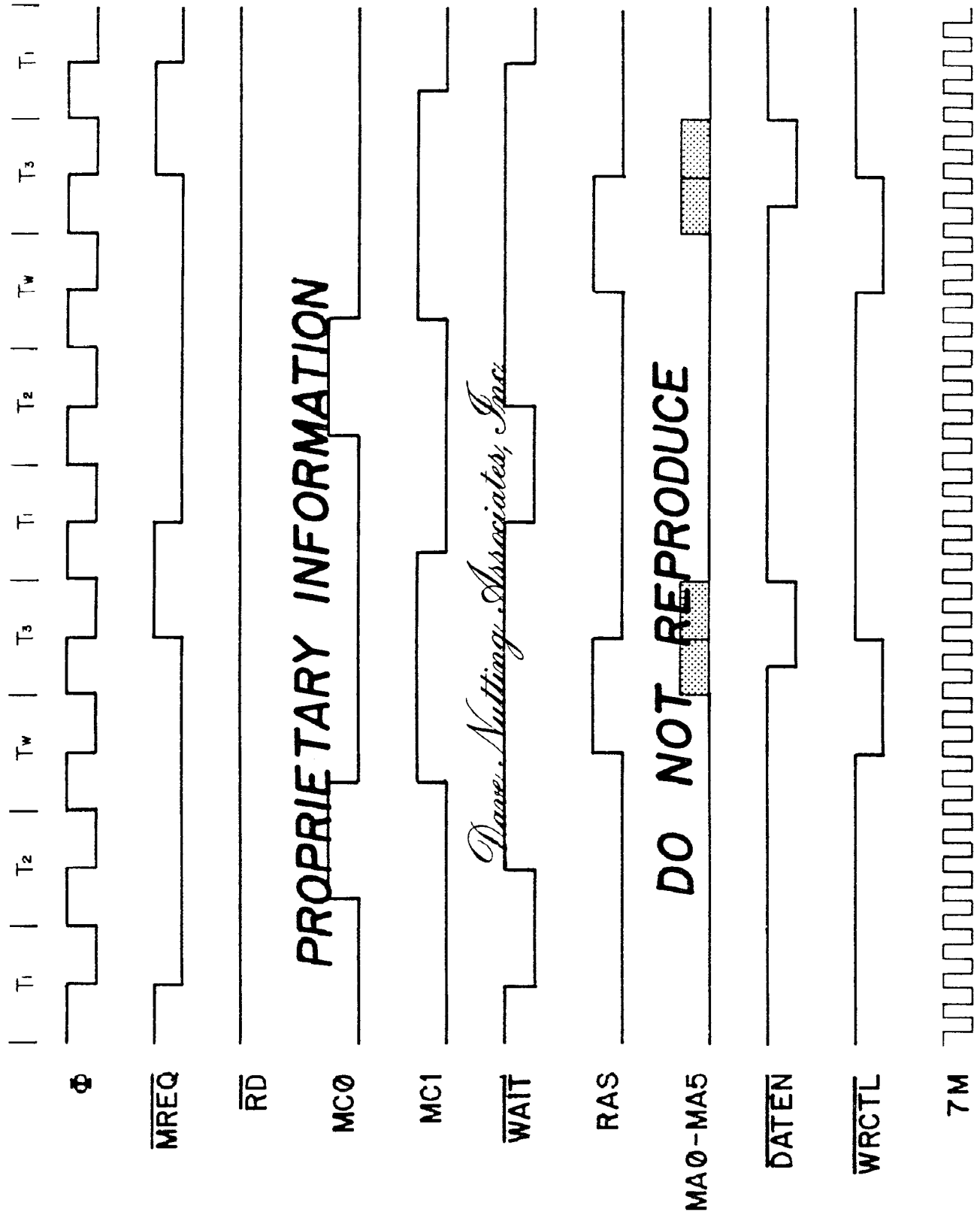
MUXD0 - MUXD7 is a 8 bit bidirectional address and data bus for the custom chips. By using this technique 16 bits of address and 8 bits of data can be sent to the custom chips on 8 wires. The state of the bus is determined by MC0 and MC1 from the data chip and RFSH from the Z80.

<u>RFSH</u>	<u>MC1</u>	<u>MC0</u>	
L	L	L	A0 - A7 to custom chips.
L	L	H	A0 - A7 to custom chips
L	H	L	A0 - A7 to custom chips
L	H	H	A0 - A7 to custom chips
H	L	L	A0 - A7 to custom chips
H	L	H	A8 - A15 to custom chips
H	H	L	D0 - D7 to custom chips
H	H	H	D0 - D7 from custom chips

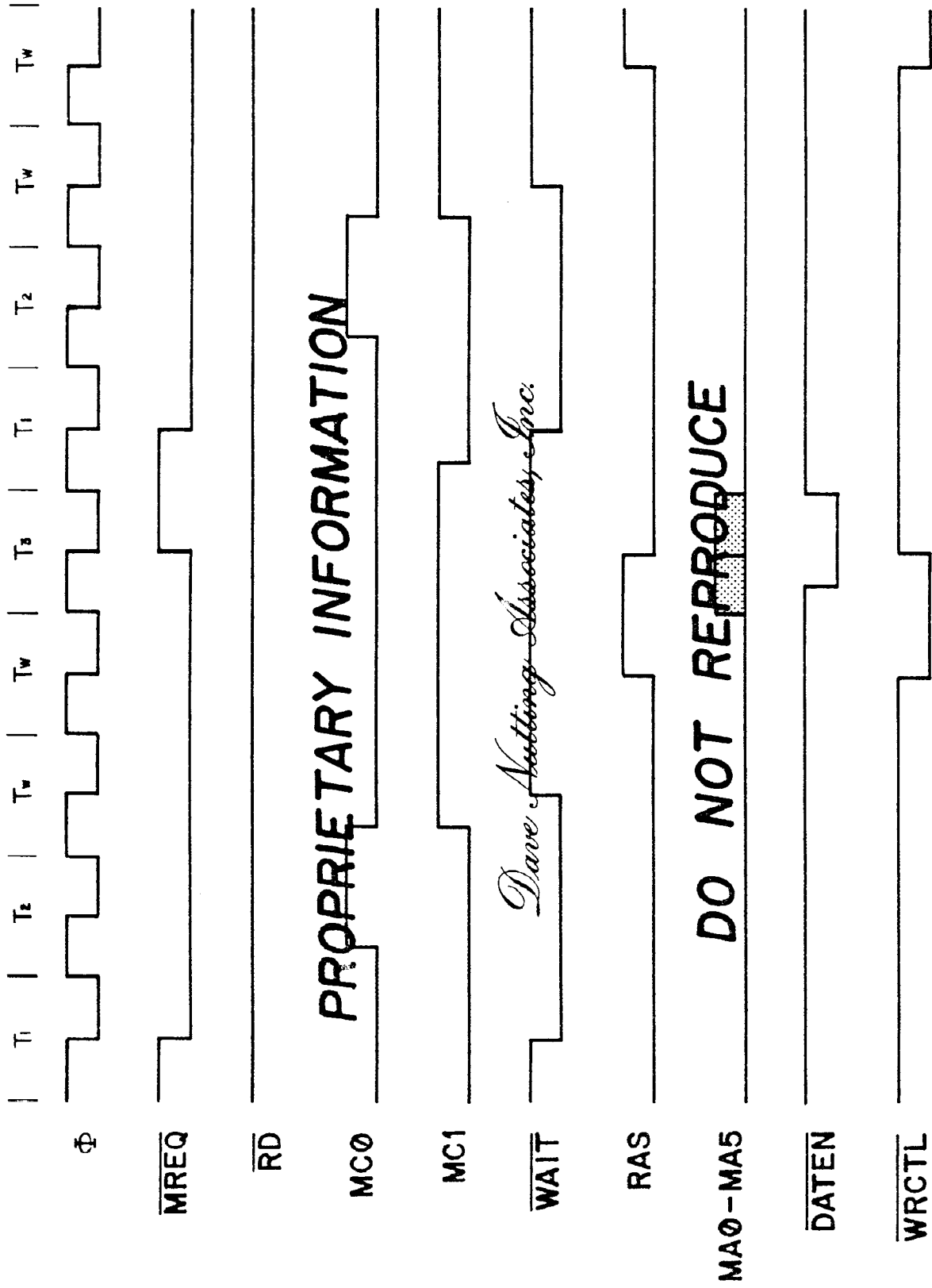
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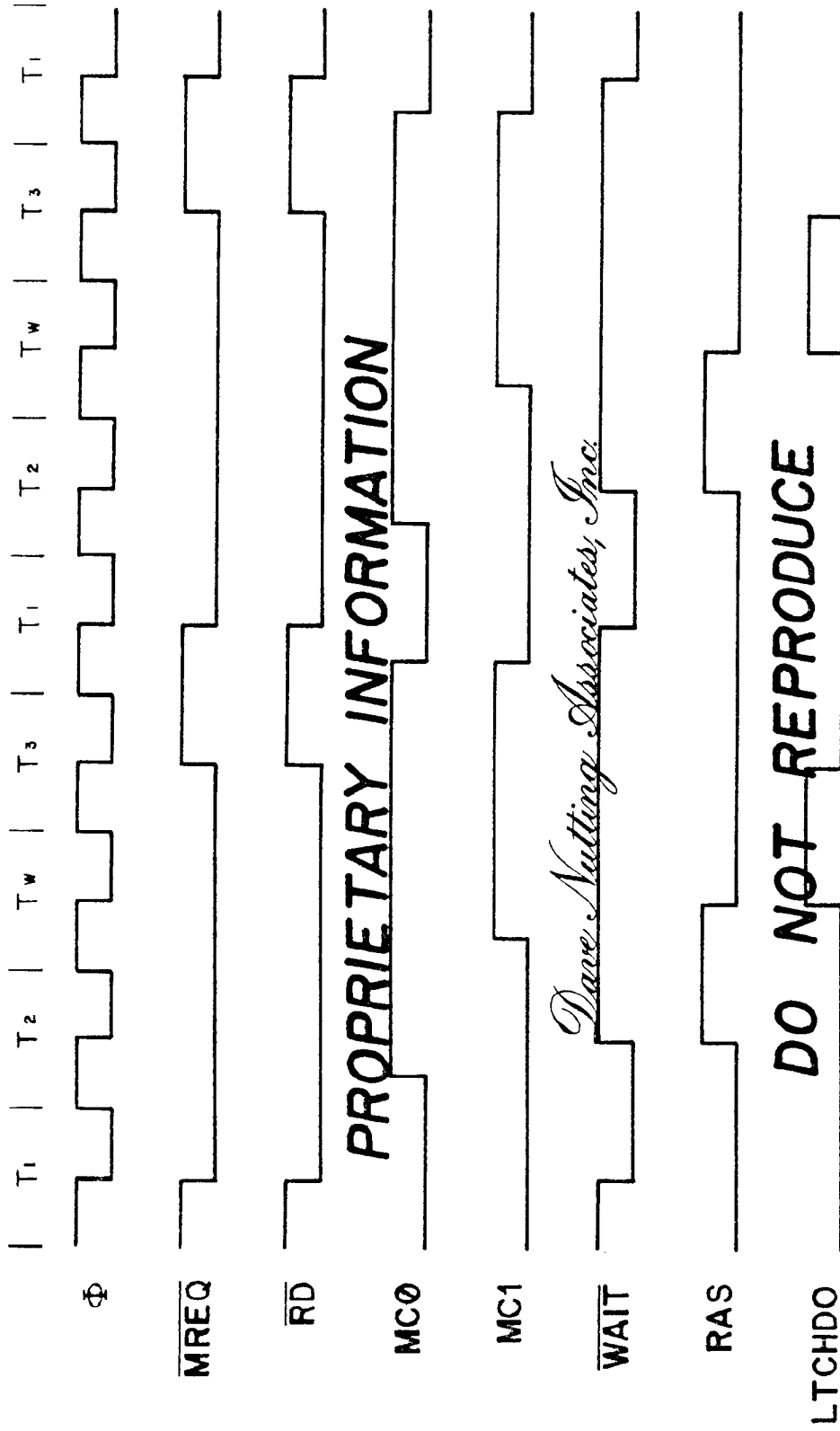
DO NOT REPRODUCE



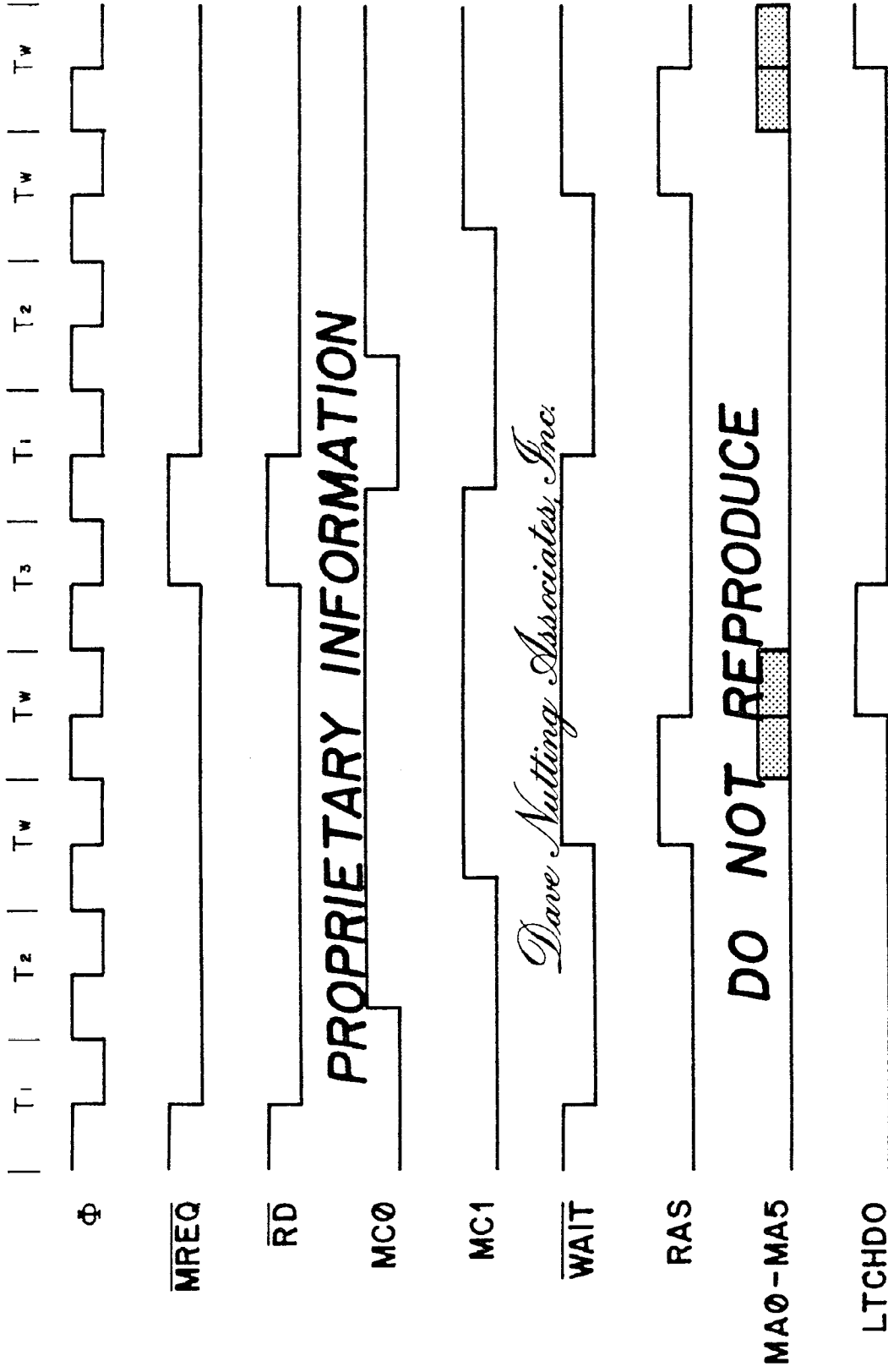
MEMORY WRITE WITHOUT EXTRA WAIT STATE



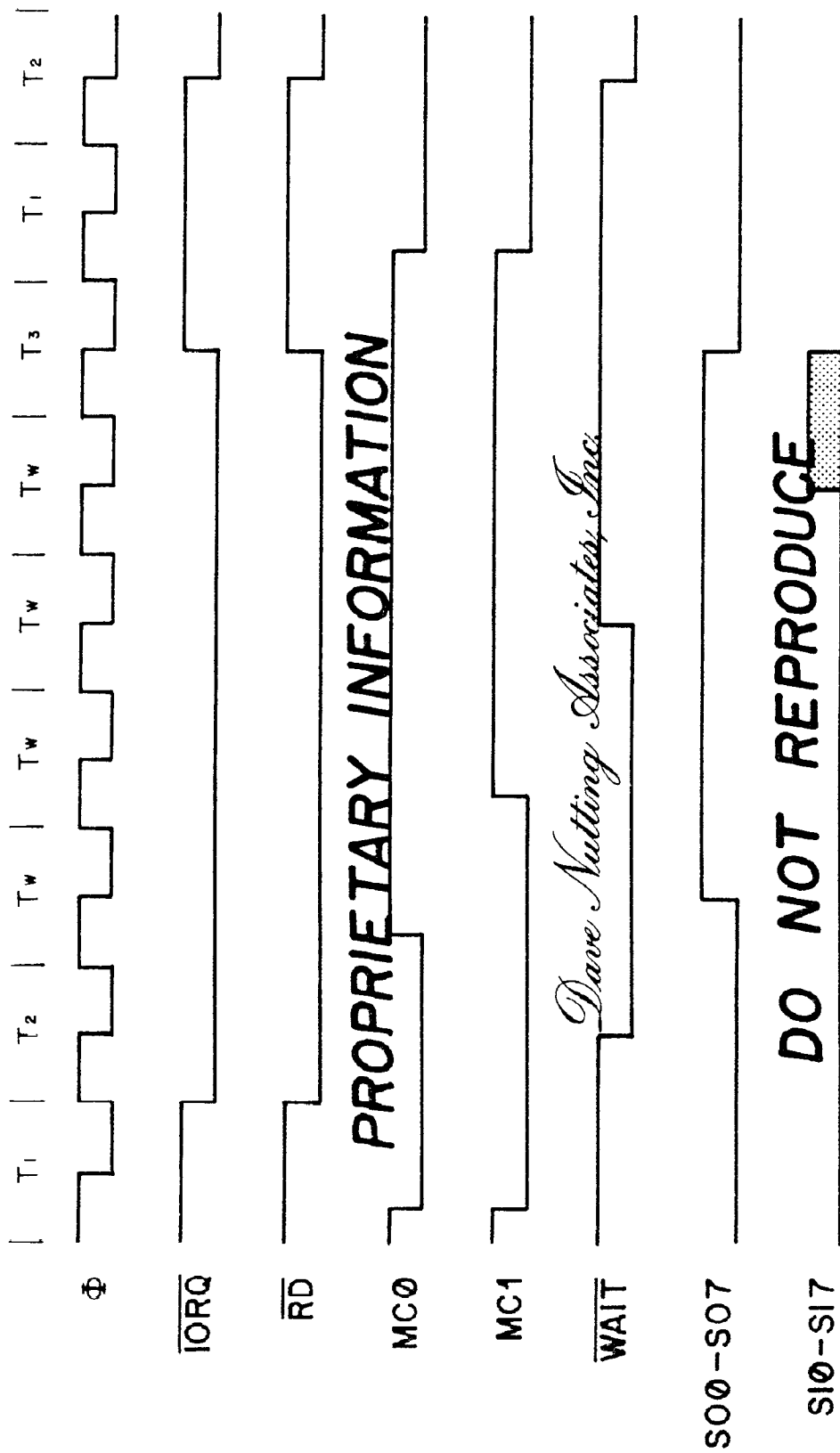
MEMORY WRITE WITH VIDEO WAIT STATE



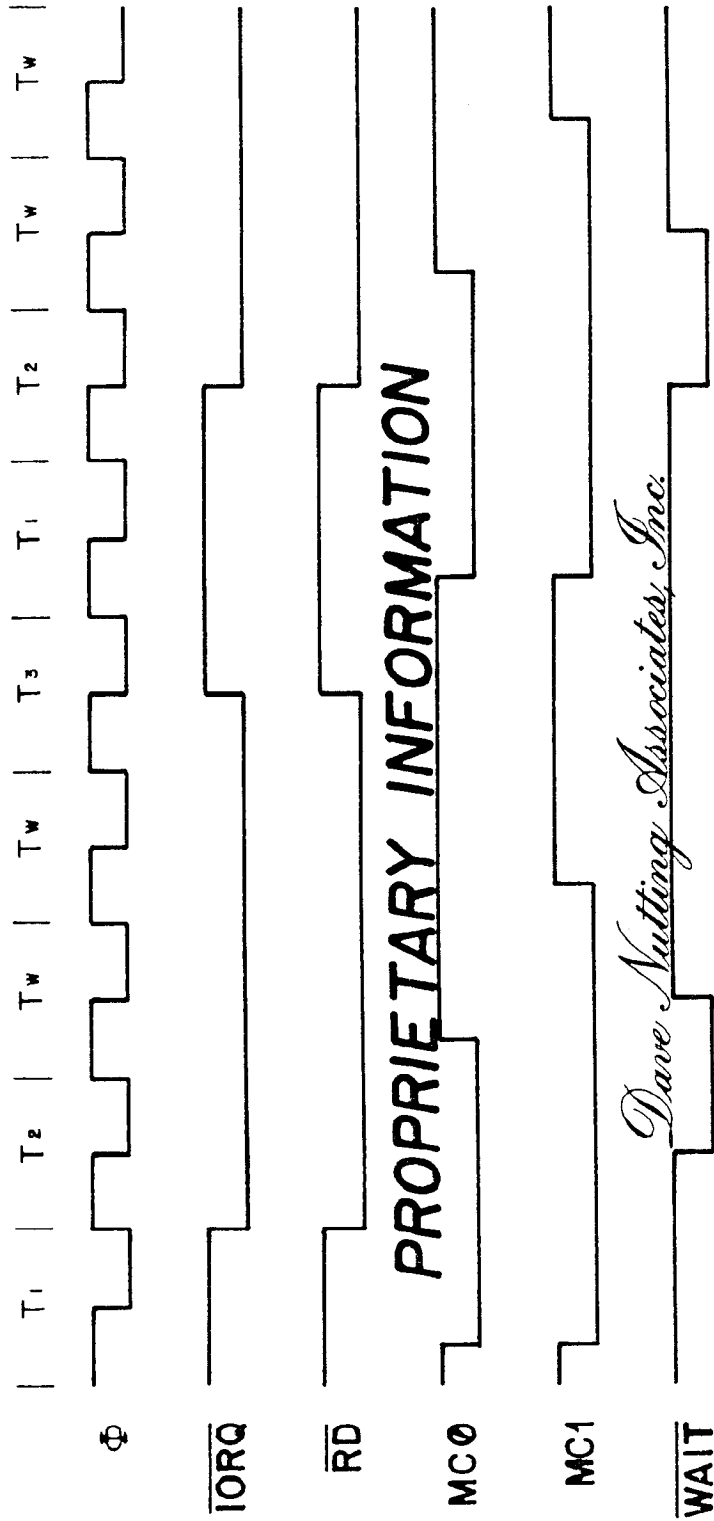
MEMORY READ WITHOUT EXTRA WAIT STATE



MEMORY READ WITH VIDEO WAIT STATE

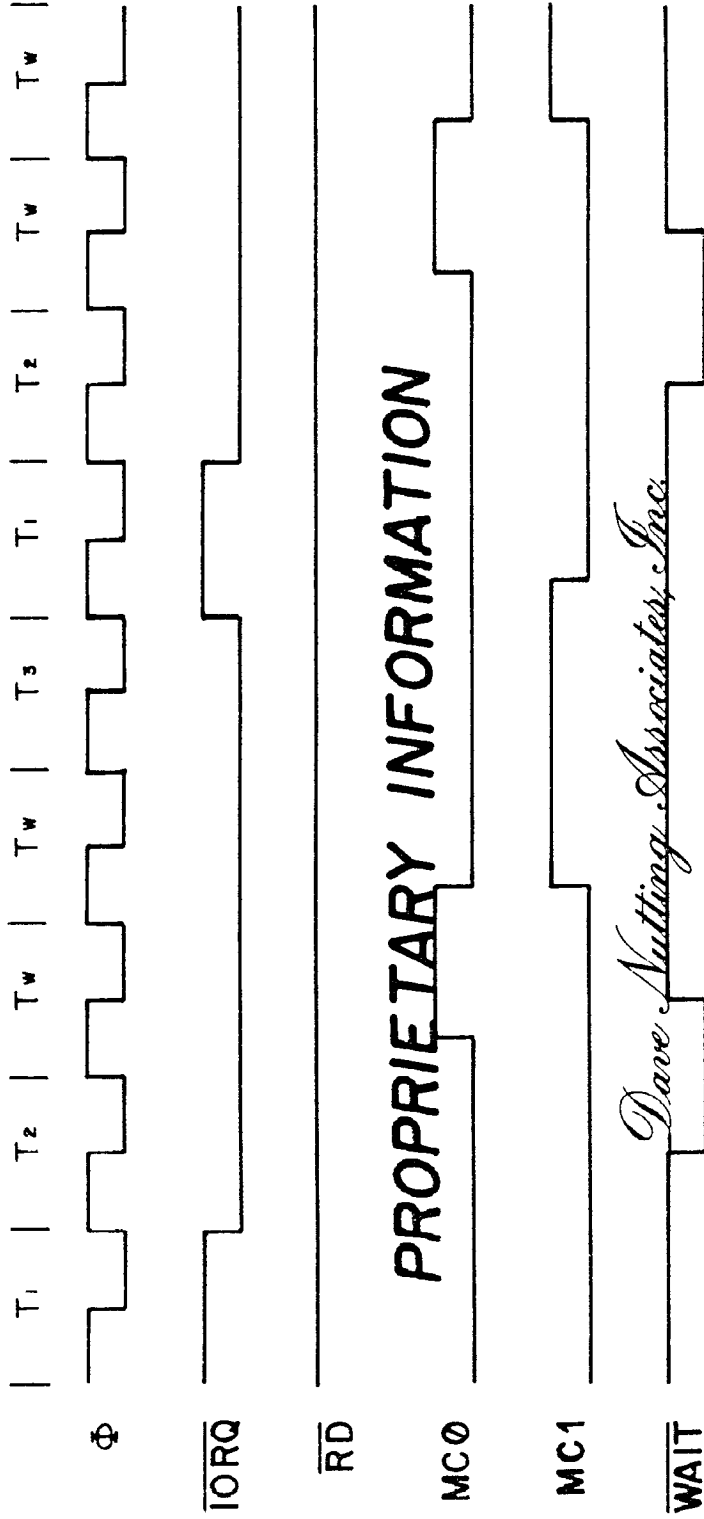


I/O READ FROM PORT 10H-17H



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I/O READ FROM OTHER THAN PORT 10H-17H



DO NOT REPRODUCE

I/O WRITE

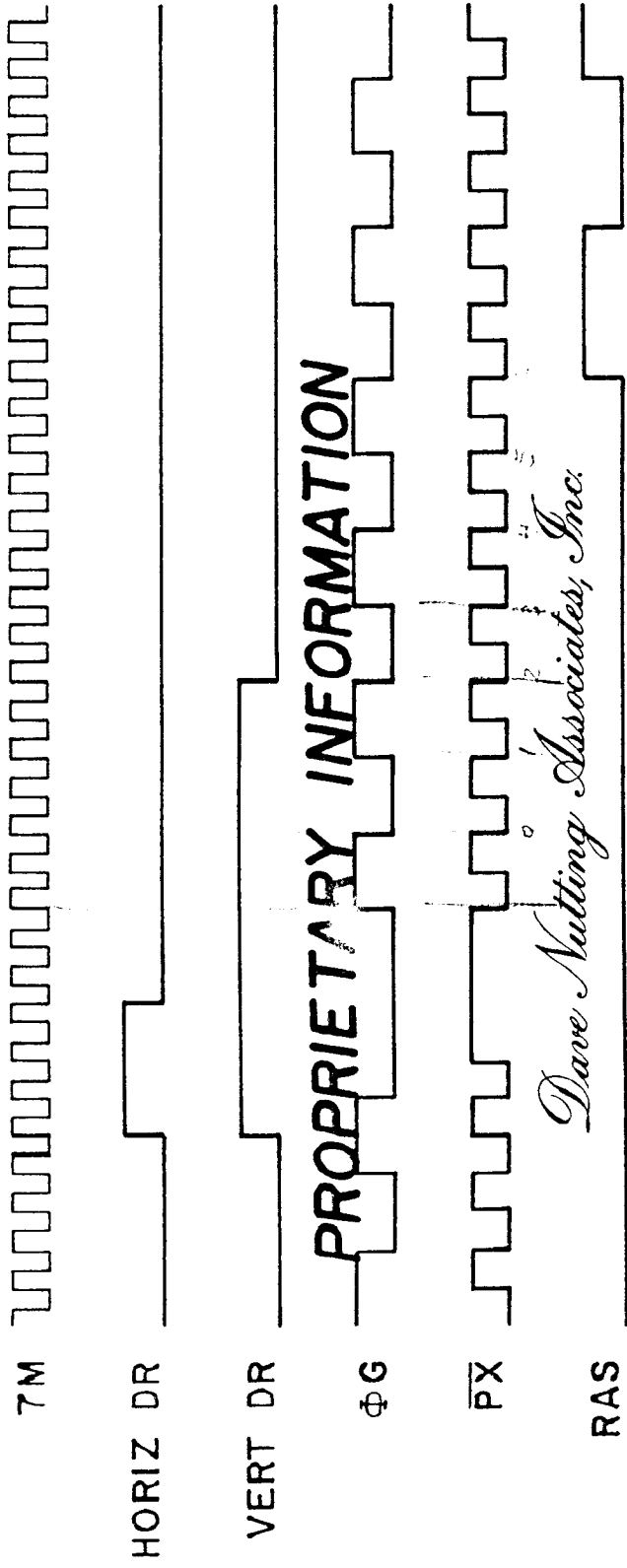
VIDEO TIMING

The frequency of \overline{PX} is half that of 7M and the \emptyset is one-fourth 7M. There are 455 cycles of 7M per horizontal line and $113 \frac{3}{4}$ \emptyset cycles per line. Because of the extra $\frac{3}{4}$ cycle \emptyset must be resynchronized at the beginning of each line. This is done by stalling \emptyset for 3 cycles of 7M. \overline{PX} is also stalled for the same amount of time. The timing relationship is shown below. The diagram also shows the relationship of VERT DR to HORIZ DR. The two RAS pulses shown are the first two video RAS signals of a line, each line contains forty.

PROPRIETARY INFORMATION

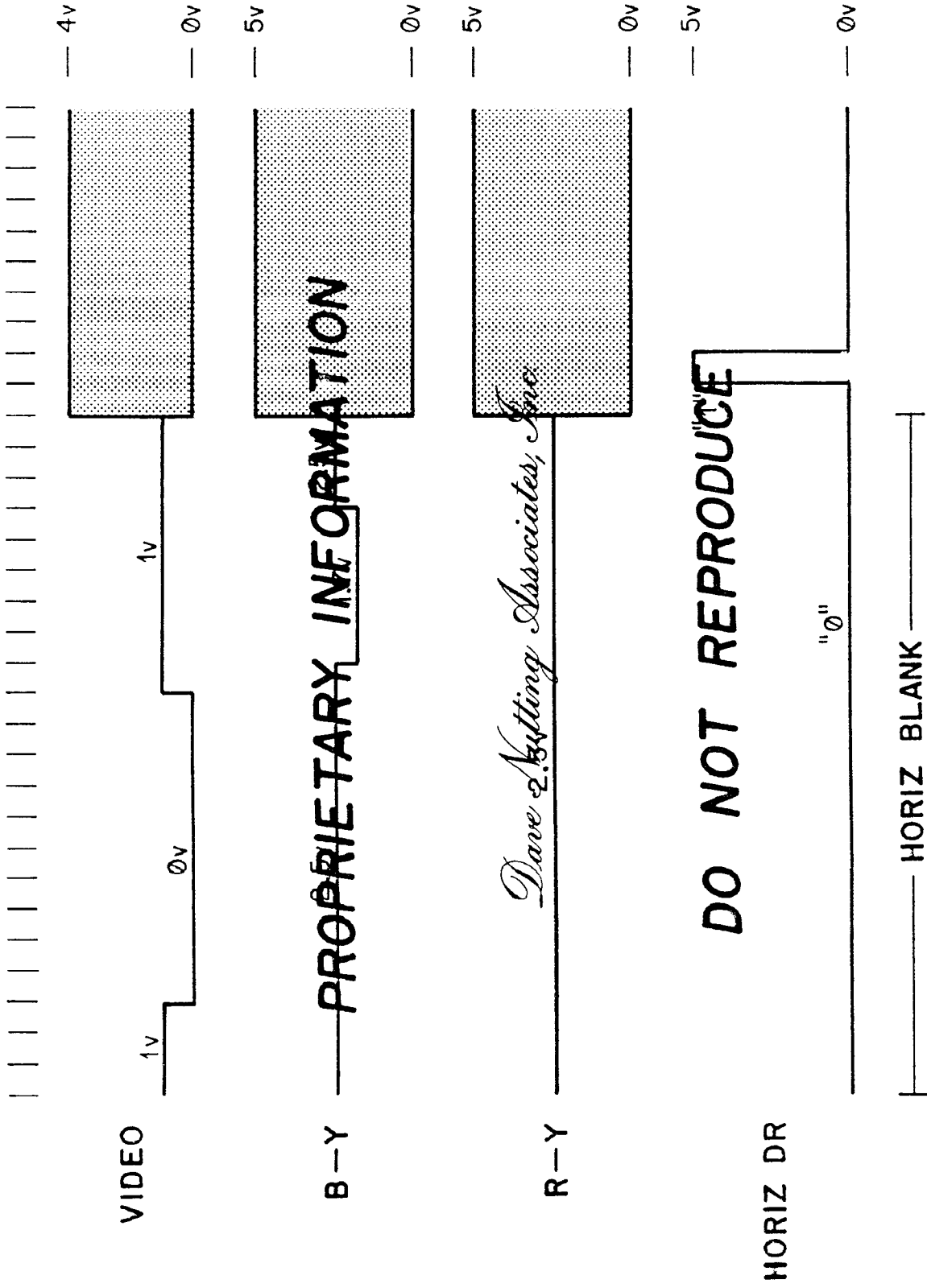
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RELATIONSHIP BETWEEN 7M, HORIZ DR, VERT DR, ΦG , \overline{PX} AND RAS

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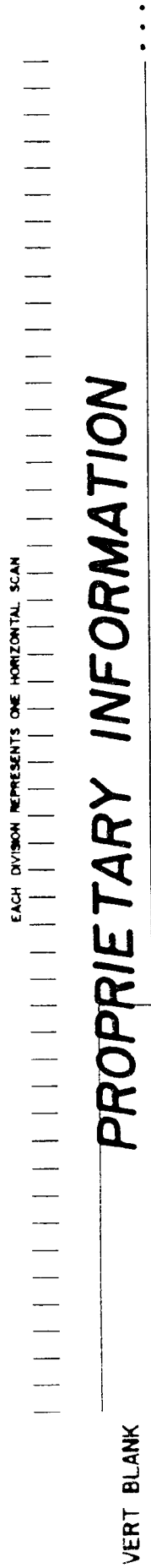


RELATIONSHIP BETWEEN HORIZ DR, HORIZ BLANK, HORIZ SYNC AND COLOR BURST

EACH HORIZONTAL DIVISION IS EQUAL TO 3 1/2 CYCLES OF 7M

THE PATTERN REPEATS EVERY 455 CYCLES OF 7M

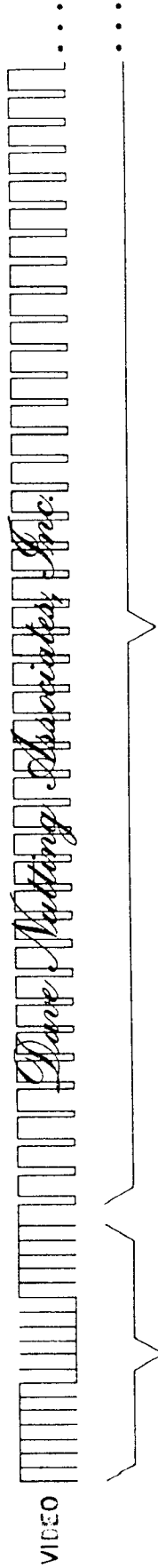
SHADED AREA VOLTAGE DETERMINED BY THE DATA IN RAM



VERT BLANK



VERT DRIVE



VIDEO

VERTICAL SYNC
WITH
EQUALIZATION PULSES

HORIZONTAL SYNC

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RELATIONSHIP BETWEEN VERTICAL SYNC, VERTICAL BLANK AND VERTICAL DRIVE
 EACH HORIZONTAL DIVISION REPRESENTS ONE HORIZONTAL SCAN

1/14/77
 1/27/77
 3/25/77
 7/6/77

N/C
 A 135
 B
 C

ELECTRICAL SPECIFICATION FOR MIDWAY CUSTOM CIRCUITS

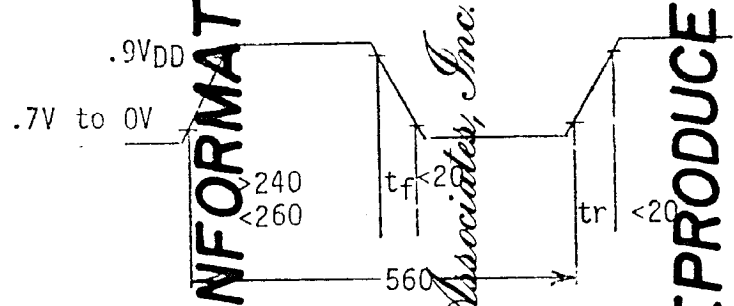
I. GENERAL SYSTEM PARAMETERS

I. A. Power Supplies

1. VDD=+5.0V $\pm 5\%$
2. VGG=+10.0V $\pm 5\%$
3. VSS=0.0V

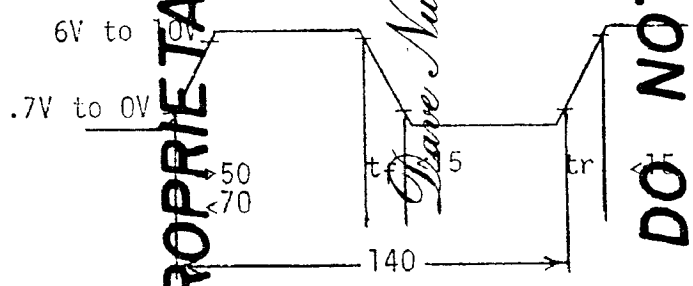
I. B. Timing Signals

1. ϕ & $\bar{\phi}$ Period = 560nsec, High time* 240nsec to 260nsec.
 ϕ and $\bar{\phi}$ have zero level crossover +1 volt -0 volts
 t_r, t_f^* less than 20nsec



(Times are in nsec)

2. $7M$ & $\bar{7M}$; Period = 140nsec, High time 50nsec to 70nsec
 $7M$ & $\bar{7M}$ have zero level crossover +1 volt -0 volt
 t_r, t_f^+ less than 15nsec



(Times are in nsec)

Dead time 5nsec
 Max C Load = 20pf

+Note
 1) High time is time clock at $\geq .6V$.
 2) Rise time from zero level to one level.

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I. B. (Continued)

*Note:

1. High time is time between 50% points.
2. Clock signals are generated by low power Shottky Logic (series 74LS). Full level swing on clock signals to be achieved through external resistor to VDD. Zero level .7V to 0V.
3. Rise time from zero level to .9VDD.

I. C. Z80 Data Bus (MUXD0-MUXD7)

1. Z80 Data Bus interface requires a three-state output/input buffer. The three states are defined below.
2. Logic 0: .5V + noise generated by chip, noise for address chip is .15V @ -430 μ A
3. Logic 1: 2.7V @ +70 μ A
4. High Impedance: Leakage at either logic 0 or 1 to be less than 5 μ A.
5. Transient Response: Transition from High Impedance to logic 0 or 1 will be complete within 442nsec of the 90% point of $\bar{\phi}$ of the last wait state of input cycle or 442nsec of the 90% point of $\bar{\phi}$ of the second wait state of the interrupt acknowledge cycle. The maximum load will be 80pf. This includes 14pfd for two custom chips.
6. Exception: The path through the Data chip connecting the RAM bus with the Z80 bus shall introduce a maximum of 160nsec of delay.
7. The low address byte will be valid on the Z80 Data Bus at least 62nsec before $\bar{\phi}$. The high address byte will be valid at least 79nsec before $\bar{\phi}$. The data byte will be valid 55nsec before $\bar{\phi}$.

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I. D. RAM Data Bus (MDO-MD7) - Home Game

1. The RAM Data Bus will require three state logic buffers.
2. Logic 0: .5V @ -25 μ A
3. Logic 1: 2.7V @ +25 μ A
4. High Impedance: 5 μ A maximum leakage at either logic 0 or 1.
5. Transient Response: The outputs shall transition from High Impedance to 0 or 1 within 120nsec of 7M. The outputs shall transition from 1 or 0 to high impedance within 20nsec of 7M. Maximum load will be 20pf.

I. E. RAM Data Bus (MDO-MD7) - Commercial Game

1. The RAM Data Bus will require three state logic buffers.
2. Logic 0: .5V @ -200 μ A
3. Logic 1: 2.7V @ +25 μ A
4. High Impedance: 5 μ A maximum leakage of either logic 0 or 1.
5. Transient Response: The output shall transition from High Impedance to 0 or 1 within 120nsec of 7M. The output shall transition from 1 or 0 to High Impedance within 2nsec of 7M. Maximum load will be 10pf.

I. F. Ambient operating temperature $\geq 0^{\circ}\text{C}$, $\leq 70^{\circ}\text{C}$.

I. G. Storage temperature $\geq -65^{\circ}\text{C}$, $\leq 150^{\circ}\text{C}$.

I. H. Packing 40 pin plastic.

II. CUSTOM CIRCUIT SPECIFICATION

This specification defines the terminal characteristics for each of the custom circuits. These specifications shall take precedence in case of conflict. All $\bar{\phi}$ references refer to the $\bar{\phi}$ and $\bar{\phi}$ inputs to the address and I/O chip.

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II. A. Data Chip

1. Input Pin List	V_0 (V)	V_1 (V)	t_d (Low) ¹ (nsec)	t_d (High) ¹ (nsec)	Ref.
<u>MREQ</u>	.5	2.45	132	6	7M
<u>RD</u>	.5	2.45	12	6	7M
<u>IORQ</u>	.5	2.45	112	126	7M
<u>7M</u>	See Section I.B.				
<u>7M</u>	"				
<u>WRCTL</u>	.5	3.1	82	82	7M
<u>MT</u>	.5	2.45	12	82	7M
<u>LTCHDO</u>	.5	3.1	120	120	7M
<u>Serial 0</u>	.5	2.45	30	30	7M
<u>Serial 1</u>	.5	2.45	30	30	7M

2. Power Supplies
See Section I. A.

3. Bus Connections

<u>MXD0</u>	See Z80 Data Bus Spec. Section I.C.
<u>MXD1</u>	"
<u>MXD2</u>	"
<u>MXD3</u>	"
<u>MXD4</u>	"
<u>MXD5</u>	"
<u>MXD6</u>	"
<u>MXD7</u>	"
<u>MD0</u>	See RAM Data Bus Spec Section I.D.
<u>MD1</u>	"
<u>MD2</u>	"
<u>MD3</u>	"
<u>MD4</u>	"
<u>MD5</u>	"
<u>MD6</u>	"
<u>MD7</u>	"

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4. Outputs	$\frac{V_O}{(V)}$	$\frac{I_O}{(\mu A)}$	$\frac{V_I}{(V)}$	$\frac{I_I}{(\mu A)}$	$\frac{CAP}{(pf)}$	$\frac{t_p}{(nsec)}$	Ref.
VIDEO*	*				10	100	7M
R-Y*	*				10	600	
B-Y*	*				10	600	
HORIZ DR	Note 4	400	2.7	20	20	20	7M
VERT DR	Note 4	400	2.7	20	20	20	7M
2.5V ⁶	--	--	--	--	--	DC	
\emptyset	Note 4	400	2.7	20	10	100	7M
PXCLK	Note 4	400	2.7	20	10	100	7M
MCO	Note 4	400	2.7	20	10	120	7M
MC1	Note 4	400	2.7	20	10	120	7M
DATEN	Note 4	400	2.7	20	10	90	7M

*Video R-Y, B-Y are analog outputs at 140nsec rate. Video, must switch from 10% to 90% of black to white in 140nsec. R-Y and B-Y transitions not to exceed .6 μ sec.

- 1 t_d (Low) and t_d (High) is maximum time in nsec except where a minimum is shown.
- 2 For IORQ RE to \emptyset t_d (Low)=132nsec t_d (High)=6nsec.
- 3 Serial 0 and Serial 1 will operate at 7MHz.
- 4 .5V + noise generated by chip.
- 5 Tap on both resistor chains for a capacitor. Will become test input with voltage applied > 8V.
- 6 The Z80 \emptyset generated by this signal with a clock driver which introduces a delay of <20nsec.

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II. B. I/O Chip

1. Input Pin List	<u>V₀</u>	<u>V₁</u>	<u>Ref</u>	<u>t_d (High)</u> (nsec)	<u>t_d (Low)</u> (nsec)
Reset	.5	2.45			
MONOS	Note 1				
RD	.5	2.45	Ø or Ø	166	172 Ø or Ø
IORQ	.5	2.45	Ø ⁶	146 Ø	132 Ø
Ø	See Section	I.B.			
Ø	"	"	"		
SIØ	.5	3.3			Note 3
SI1	.5	3.3			Note 3
SI2	.5	3.3			Note 3
SI3	.5	3.3			Note 3
SI4	.5	3.3			Note 3
SI5	.5	3.3			Note 3
SI6	.5	3.3			Note 3
SI7	.5	3.3			Note 3
TEST		5.0			DC
2. Power Supplies					
See Section I.A.					
3. Bus Connections					
MUXDØ	See Z80 Data	Bus Spec	Section I.C.		
MUXD1	"		"		
MUXD2	"		"		
MUXD3	"		"		
MUXD4	"		"		
MUXD5	"		"		
MUXD6	"		"		
MUXD7	"		"		
4. Outputs					
	<u>V₀</u> (V)	<u>I₀</u> (µA)	<u>V₁</u> (V)	<u>I₁</u> (µA)	
Audio	Note 4	Fmax - 20KHz			
Discharge	Note 5	.5V 4V			
SØØ	Note 3	Note 7 200	4V	1650	
SØ1	Note 3	Note 7 200	4V	1650	
SØ2	Note 3	Note 7 200	4V	1650	
SØ3	Note 3	Note 7 200	4V	1650	
SØ4	Note 3	Note 7 200	4V	1650	
SØ5	Note 3	Note 7 200	4V	1650	
SØ6	Note 3	Note 7 200	4V	1650	
SØ7	Note 3	Note 7 200	4V	1650	
POT Ø	Note 2	5	VDD-.5	50	
POT 1	Note 2	5	VDD-.5	50	
POT 2	Note 2	5	VDD-.5	50	
POT 3	Note 2	5	VDD-.5	50	

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Note 1 MONOS triggers at 2.1 volts $\pm 2\%$ \pm noise voltage when the supply is 5.25V.

Note 2 Open source-Voltage measured with 0.2ma.

Note 3 Time from load of address into microcycle register to data valid on MUX data bus from SI inputs (data path through address decoder, out on S0 outputs, through closed switch and isolation diode, into SI input to MUX Data Bus) shall be 2 μ sec max. Drop of isolation diode will be 0.7V max. S0 must drive 2k Ω in the high level. Max C load of S0 shall be 300 pf. SI input shall have kill device enabled by INPUT.

Note 4 Audio voltage oscillates between 0V and one of the following voltages; .33, .67, 1.00, 1.33, 1.67, 2.00, 2.33, 2.67, 3.00, 3.33, 3.67, 4.00, 4.33, 4.67 and 5.00. These voltages should be $\pm 6\%$. The load shall be 1000pf and 100k Ω .

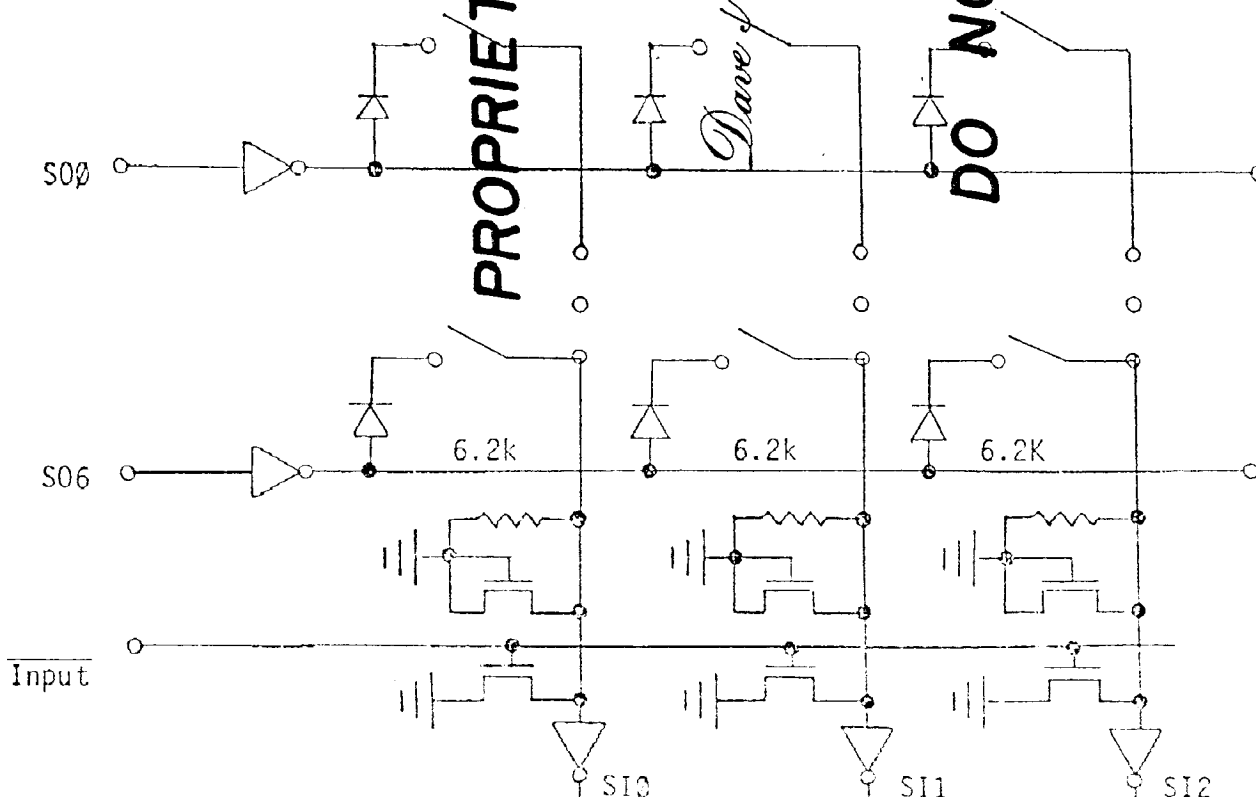
Note 5 Discharge is open drain to V_{cc} Discharges .01 μ fd capacitor to .2 τ in 144 μ sec.

Note 6 For IOREQ Ref. t_d (Low) = 2nsec t_d (High) = 166nsec.

Note 7 .5V + noise generated by I/O chip.

Miscellaneous Timing

Time for Address - 20 ma



No more than three switches on each S0 are closed at one time.

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II. C. Address Chip

1. Input Pin List

	V0 (V)	V1 (V)	t _{pd} (Low) (nsec)	t _{pd} (High) (nsec)	REF
RFSH	.5	2.45	222 \emptyset	216	\emptyset
MREQ	.5	2.45	152 \emptyset	166	\emptyset or $\overline{\emptyset}$
RD	.5	2.45	172 \emptyset or $\overline{\emptyset}$	166	\emptyset or $\overline{\emptyset}$
MI	.5	2.45	176 \emptyset	242	\emptyset
A12 ¹	.5	2.45			\emptyset
A13 ¹	.5	2.45			\emptyset
A14 ¹	.5	2.45			\emptyset
A15 ¹	.5	2.45			\emptyset
IORQ	.5	2.45	132 \emptyset	146	\emptyset ²
LIGHT	.5	2.45	Asyn		
TEST	.5	5.0	DC		
HORIZ. DR.	.5	2.45	Note 3		$\overline{\emptyset}$
VERT. DR.	.5	2.45	Note 4		\emptyset
\emptyset					
\emptyset					

2. Power Supplies

See Section I.A.

3. Bus Connections

MXD0	See Z80 Data Bus Spec Section I.E.
MXD1	"
MXD2	"
MXD3	"
MXD4	"
MXD5	"
MXD6	"
MXD7	"

4. Outputs

	V0 (V)	I0 (μ A)	V1 (V)	I1 (μ A)	CAP (pf)	t _{pd} (Low) (nsec)	t _{pd} (High) (nsec)	REF
LATCHDO	Note 7	Note 6	3.1	Note 6	10	280	140	$\overline{\emptyset}$ ⁵
WAIT	"	"	400	2.4	25	490	490	$\overline{\emptyset}$
MAO-MA5	"	"	400	2.4	20	242	240	\emptyset or $\overline{\emptyset}$
INT	"	"	400	2.4	25	490	572	\emptyset
RASO-RAS3	"	"	400	2.4	20	382	382	$\overline{\emptyset}$
WRCTL	"	"	Note 6	3.1	Note 6	10	382	\emptyset

- Time from High Impedance to 1 or 0 is 200nsec. (from \emptyset_1 of T₁)
- For IORQ Ref to \emptyset t_d (Low)=152nsec t_d (High)=166nsec. $\overline{\emptyset}$
- Horizontal Drive time from low to high is 40nsec after $\overline{\emptyset}$.
Time from high to low is 100nsec before rising edge of \emptyset .
- Vertical Drive will transition from low to high 40nsec after falling edge of \emptyset . Its width will be 2.1 μ sec max. 1.54 μ sec min. It will go from high to low 100nsec before falling edge of \emptyset .
- Reference t_{pd} (High) is \emptyset .
- MOS to MOS signal.
- .5V + noise generated by Address Chip (15V) = 65V

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See Section I.E.

III. I/O MODE DECODE

I/O Parts

<u>HEX</u>	<u>Out</u>	<u>Input</u>
0	Color 0 Right	
1	" 1 "	
2	" 2 "	
3	" 3 "	
4	" 0 Left	
5	" 1 "	
6	" 2 "	
7	" 3 "	
8	Consumer/Commercial	Intercept Feedback
9	Horiz Color Bndry	
A	Vertical Blank	
B	Color Block T	
C	Logic Reg	
D	Interrupt Feedback	
E	Interrupt Mode	Vertical Addr Feedback
F	Interrupt Line	Horizontal Addr Feedback
10	Tone Master OSC	SW Bank 0
11	Tone A	1
12	" B	2
13	" C	3
14	Tremello	4
15	Tone C Volume	5
16	Tone A,B Volume	6
17	Misc Volume	7
18	Sound Block T	
19		
1A		
1B		
1C		
1D		POT 0
1E		" 1
1F		" 2
20		" 3
21		" 4
22		" 5
23		" 6
24		" 7
.		
.		
2F		

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